Minimal Generalized Time-Bandwidth Product Method for Estimating the Optimum Fractional Fourier Order

Lin Tian¹², Zhenming Peng¹

¹School of Optoelectronic Information, University of Electronic Science and Technology of China, Chengdu, China
²Department of Electronics and Information Engineering, Yili Normal University, Yili, China
Email: tianlin20110501@163.com, zmpeng@uestc.edu.cn

Received November 2014

Copyright © 2015 by authors and Scientific Research Publishing Inc.
This work is licensed under the Creative Commons Attribution International License (CC BY).
http://creativecommons.org/licenses/by/4.0/

Abstract

A minimal generalized time-bandwidth product-based coarse-to-fine strategy is proposed with one novel ideas highlighted: adopting a coarse-to-fine strategy to speed up the searching process. The simulation results on synthetic and real signals show the validity of the proposed method.

Keywords

Generalized Time-Bandwidth Product, Coarse-to-Fine Strategy, Optimum Fractional Fourier Order, Fractional Fourier Transform

1. Introduction

To get a high resolution time-frequency distribution, fractional Fourier transform (FrFT) is a useful tool. Traditional time-frequency method combined FrFT, and fractional Gabor transform was proposed [1]-[3], the optimum order of FrFT is major factor of the transform, for it determines the appropriate fractional domain where the signal is best concentrated. In [2] the optimum order of FrFT is chosen by the chirp rate of the signal. This transform can get high resolutions, yet the time-frequency distributions are in time-fractional plane, the physics meaning is not clean. In [4] Wigner distribution function and its FrFT are used to get high resolutions results, using FrFT, and inverse FrFT, the physics meaning is clean, and Wigner distribution function and its FrFT were used to estimate the optimum order, yet the method is not appropriate for a lot of time-frequency components. The fractional optimal short-time Fourier transform was proposed [5], this method utilizes the FrFT to improve the resolution of the time-frequency distributions, and the generalized time-bandwidth product (GTPB) was
proposed. In [6] the fractional Gabor transform was proposed which has clean physics meaning, to save computational cost, the maximal amplitude in fractional domain was used for optimum fractional Fourier order in the paper. The fast algorithm of maximal amplitude has established in fractional domain [7]. For FrFT can be interpreted as a decomposition of the signal in terms of chirps, this method is appropriate for signal detection [8]. For non-stationary signal time-frequency analysis, the optimum order of FrFT can be found the minimal generalized time-bandwidth product. This method is suitable for signal which has many components or has continuous spectra. However, the minimal generalized time-bandwidth product can give the directly optimum order. Since there is no fast algorithm, this paper uses coarse-to-fine strategy, and gives a fast algorithm. Though, theoretically, the amplitudes-based method is fast, it is powerless for the signals to consist many components in time-frequency analysis. We propose a minimal-generalized time-bandwidth product-based coarse-to-fine algorithm to estimate the compact fractional domain in this paper.

2. Minimal Generalized Time-Bandwidth Product Method

2.1. Generalized Time-Bandwidth Product

Fractional Fourier transform (FrFT) is a generalization of the Fourier transform, FrFT is a linear operator. The FrFT of signal can be interpreted as the rotating the signal in the time-frequency plane, the FrFT of \( s(t) \) is defined as [9]:

\[
s_p(u) = \text{FrFT}_p(s(t)) = \int_{-\infty}^{\infty} s(t) K_p(t,u) dt
\]

where \( s_p \) is the \( p \)-th order FrFT of \( s(t) \), \( p \) is the transform order of FrFT, the period of \( p \) is 4. FrFT\(_p\) is an operator, means the \( r \)-th order FrFT. \( K_p(t,u) \) is the transform kernel which is given as follows [10]:

\[
A_\emptyset = e^{-\frac{\pi}{4} \frac{\text{sgn} \sin(\emptyset)}{\sin^2(\emptyset)} - j \frac{\emptyset}{2}}
\]

\[
K_p(t,u) = A_\emptyset \exp\left[j\pi t^2 \cot(\emptyset) - 2tu \csc(\emptyset) + u^2 \cot(\emptyset)\right]
\]

\[
\emptyset = \frac{\pi}{2} p
\]

In light of the properties of the FrFT, the interval of the optimum order can be restricted in [0, 2) [11]. The time-bandwidth product (TBP) of the signal \( s(t) \) of the weighted window function is defined as follows [6]:

\[
\text{TBP} = T_{s,h} * B_{s,h}
\]

\( T_{s,h} \) and \( B_{s,h} \) represent the time and the frequency width of the signal of the weighted window function, respectively.

If the \( p \)-th order FrFT of signal \( s(t) \) is \( s_p(u) \), and the \( p \)-th order FrFT of window function is \( h_p \). The GTPB can be written as

\[
\text{GTBP} = T_{s,p} * B_{s,p}
\]

where \( T_{s,p} \) is the time length of signal \( s_p \) and the window function, and \( B_{s,p} \) is the frequency width of signal \( s_p \) and the window function.

\[
T_{s,p} = \left(T_{s,p}^2 + T_{h,p}^2\right)^{1/2}
\]

where \( T_{s,p} \) is the time length of signal \( s_p \), \( T_{h,p} \) is the time length of window function \( h_p \), and the \( T_{s,p} \) can be given by:

\[
T_{s,p} = \left[\int (u - \varphi)^2 |s_p(u)|^2 du\right]^{1/2}
\]

\[
\|f_p\|
\]
where \( \varphi_u \) is the time length center, \( \| s_p \| \) is the 2-norm of \( s_p \), and \( \varphi_u \) can be gotten by:

\[
\varphi_u = \frac{\int [u|s_p(u)|^2 \, du]^{1/2}}{\| s_p \|}
\]

(7)

\( T_{sp} \) can be gotten by the same method as \( T_{sp} \).

\( B_{s_p, h_p} \) can be given by:

\[
B_{s_p, h_p} = \left( B_{s_p}^2 + B_{h_p}^2 \right)^{1/2}
\]

(8)

where \( B_{s_p} \) is the bandwidth of \( s_p \), \( B_{h_p} \) is the bandwidth of \( h_p \). \( B_{s_p} \) can be given by:

\[
B_{s_p} = \left[ \frac{\int (\omega - \varphi_\omega)^2 |S_p(\omega)|^2 \, d\omega}{\| S_p(\omega) \|} \right]^{1/2}
\]

(9)

where \( S_p(\omega) \) is the Fourier transform of \( s_p \), and \( \varphi_\omega \) is the frequency center of \( S_p(\omega) \). \( \varphi_\omega \) can be given by:

\[
\varphi_\omega = \frac{\int \omega |S_p(\omega)|^2 \, d\omega}{\| S_p(\omega) \|}^{1/2}
\]

(10)

\( B_{h_p} \) can be gotten by the same method as \( B_{s_p} \).

### 2.2. Optimum Order of FrFT Estimated by GTBP

The optimum order of FrFT can be given by minimal generalized time-bandwidth product:

\[
p_{opt} = \arg \min_{p \in [4]} \{ \text{GTBP} \}
\]

(11)

According to the operation properties of FrFT, the order of FrFT can be narrowed to the range of \( 0 \leq p < 2 \) [11]:

\[
p_{opt} = \arg \min_{p \in [0,2]} \{ \text{GTBP} \}
\]

(12)

### 2.3. Algorithm of MGTBP

Suppose \( \Delta p, \alpha, \epsilon \) are constants greater than 0 and less than 1, the MGTBP algorithm can be implemented by the following steps:

Step 1: \( p_{\text{begin}} = 0 \), \( p_{\text{end}} = 2 \), \( \Delta p = \Delta p_0 \).

Step 2: let \( p = \{ p_{\text{begin}}, \Delta p + p_{\text{begin}}, 2\Delta p + p_{\text{begin}}, \ldots, p_{\text{end}} = 2 \} \).

Step 3: perform FrFT of the signal for each \( p \).

Step 4: Find \( p_{opt} \) with (4) and (12).

Step 5: \( p_{\text{begin}} = \max \{ 0, p_{opt} - 0.5\Delta p \} \), \( p_{\text{end}} = \min \{ 2, p_{opt} + 0.5\Delta p \} \), \( \Delta p = \alpha \Delta p \), go to Step 2, until \( \Delta p < \epsilon \).

The above loops can be stopped until the number of cycles equals to the number \( M \):

\[
M = \left[ \frac{\lg(\frac{\Delta p}{\epsilon})}{\lg(1/\alpha)} \right]
\]

(13)

where \( \left[ \ldots \right] \) is the upward integral function.
Figure 1 shows the main idea of such coarse-to-fine of the general time-bandwidth product (GTBP). At the beginning, GTBP is computed for different values of order of FrFT that are distributed over the interval [0, 2], and the optimum order is roughly found. Then a finer searching will be done on a small interval around the optimum order. This procedure will be repeated until the obtained satisfies the predefined accuracy requirement.

2.4. Computational Complexity of MGTBP

Suppose an N points signal $s(t)$ and N points signal window function $h$ for the given order of windowed signal, to get $S_p$ and $h_p$, the complexity of FrFT is $O(2N\log N)$ [4], the complexity of Fourier transform of $S_p$ and $h_p$ is $O(2N\log N)$, and the complexity of per loop iteration is $O(4N\log N)$. The whole searching processing computational complexity is $O(4MN\log N)$.

3. Simulation Results and Analysis

The MGTBP is implemented for searching the optimum order of FrFT, the fast discrete FrFT algorithm is reported by Ozakatas et al. [10]. The method proposed in our paper is compared with the MACF method [8]. First, A signal is used to evaluate the synthetic signal consisting of two attenuated chirp components, the signal can be written as the following formula:

$$x(t) = 4\exp\left(j\pi(t^2 + 4t)\right)\exp\left(-\frac{\pi}{5}t^2\right) + \exp\left(j\pi(0.5t^2 - 2.5t)\right)\exp\left(-\frac{\pi}{9}t^2\right).$$

The parameter settings take the values as follows: $\epsilon = 0.001$, $\alpha = 0.5, \Delta p = 0.1$. The MACF method gives 0-th order as optimum fractional Fourier order, and MGTBP estimates 0.737-th order as the optimum fractional Fourier order. The fractional Gabor transforms in different optimum orders give as Figure 2. Figure 2(a) shows the 0-th order fractional Gabor transform, the optimum fractional Fourier order is given by MACF method. Figure 2(b) shows the 0.737-th order fractional Gabor transform, the optimum fractional Fourier order is given by MGTBP method. The above results imply, although the MGTBP method has a more complicated searching technique, the MGTBP has a better ability to find the compact domain for time-frequency analysis.

4. Discussion and Conclusions

In this paper, the MGTBP method for estimating the optimum fractional Fourier order has been developed. Its computational complexity is $O(4MN\log N)$, although the computational complexity is $O(MN\log N)$ [8]. The searching algorithm adopted by the MACF is simpler than MGTBP, generally speaking, the optimum fractional Fourier order searching by MGTBP is more appropriate for time-frequency analysis than MACF. Further, our simulation results on the synthetic signal shows that the MGTBP has better performance on finding optimal fractional domain.

Figure 1. Schematic showing the stream of the coarse-to-fine procedure.
Figure 2. Fractional Gabor transform at different optimum order. (a) 0-th order as optimum order given by MACF; (b) 0.737-th order as optimum order given by MGTBP.

Acknowledgements
The authors wish to thank the National Natural Science Foundation of China (Grants No. 41274127) and the Yili Normal University research project (Grants No. 2014YSYB04) for financial support of this research.

References


